Joint Channel Estimation and Turbo Receiver

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Benchmarking Capabilities of Evolutionary Algorithms in Joint Channel Estimation and Turbo Multi-User Detection/Decoding

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## Motivations

- What critical to a communication signal processing application are performance and complexity
  - Optimal solutions are ofter NP-hard to obtain, with unaffordable cost
  - Traditionally, suboptimal solutions are sought, at lower complexity
- Evolutionary algorithms are capable of offering near optimal performance with affordable cost
  - A well-tuned EA may solve a NP-hard problem with complexity at most polynomial in problem size
- We evaluate several evolutionary algorithms in a very challenging application
  - Joint channel estimation and turbo multiuser detection-decoding for OFDM

## Background

- Joint channel estimation and turbo multiuser detection-decoding
  - Turbo MUD/decoding optimisation given CSI is NP-hard, and optimal ML solution is computationally prohibitive
- Within joint optimisation of iterative CE and MUD/decoding
  - CE optimisation is defined on continuous space while MUD optimisation is defined on discrete space
- We test both discrete-binary and continuous-valued
  - Genetic algorithm, repeated weighted boosting search, particle swarm optimisation, differential evolution algorithm
- EA aided joint CE and turbo multiuser detector/decoder:
  - BER approaches ML bound associated with perfect CSI
  - CE accuracy attains optimal Cramér-Rao lower bound
  - Complexity is a fraction of NP-hard optimal ML complexity

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## System Schematic



- SDMA: U single-antenna users are spatially separated by user-specific CIRs
- OFDM: K subcarriers for combating dispersive channel
- BS: has Q antennas and performs soft-in soft-out iterative detection and decoding

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## Joint CE and MUD

Joint ML CE and MUD solution

$$ig(\widehat{\pmb{h}}[\pmb{s}],\widehat{\pmb{X}}[\pmb{s}]ig) = rg\min_{\pmb{h}[\pmb{s}],\pmb{X}[\pmb{s}]} Jig(\pmb{h}[\pmb{s}],\pmb{X}[\pmb{s}]ig)$$

with joint optimisation cost function

$$J(\boldsymbol{h}[\boldsymbol{s}], \boldsymbol{X}[\boldsymbol{s}]) = \sum_{q=1}^{Q} \|\boldsymbol{Y}_{q}[\boldsymbol{s}] - \boldsymbol{X}^{\mathrm{T}}[\boldsymbol{s}]\overline{\boldsymbol{\mathsf{F}}}\boldsymbol{\mathsf{h}}_{q}[\boldsymbol{s}]\|^{2}$$

- Y<sub>q</sub>[s] ∈ C<sup>K×1</sup>: qth antenna received data over K subcarriers at sth OFDM symbol
- X[s] = [X<sup>1</sup>[s] X<sup>2</sup>[s] ··· X<sup>U</sup>[s]]<sup>T</sup> with X<sup>u</sup>[s] = diag{X<sup>u</sup>[s, 1], ··· , X<sup>u</sup>[s, K]} being transmitted data of user *u* over *K* subcarriers at *s*th OFDM symbol

• 
$$\overline{\mathbf{F}} = \text{diag}\{\underbrace{\mathbf{F}, \mathbf{F}, \cdots \mathbf{F}}_{U}\}$$
 with  $\mathbf{F} \in \mathbb{C}^{K \times L_{cir}}$  denoting FFT matrix

h<sub>q</sub>[s] = [(h<sub>q</sub><sup>1</sup>[s])<sup>T</sup> (h<sub>q</sub><sup>2</sup>[s])<sup>T</sup> · · · (h<sub>q</sub><sup>U</sup>[s])<sup>T</sup>]<sup>T</sup> with h<sub>q</sub><sup>U</sup>[s] ∈ ℂ<sup>L<sub>cir</sub>×1</sup> being CIR vector between *u*th user and *q*th receive antenna during *s*th OFDM symbol

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#### Given CSI, ML MUD solution

$$\widehat{\mathbf{X}}[s,k] = \arg\min_{\mathbf{X}[s,k]\in\mathcal{S}^U} J_{mud}(\mathbf{X}[s,k]), 1 \le k \le K$$

with symbol set  $\ensuremath{\mathcal{S}}$  and MUD optimisation cost function

$$J_{mud}(\mathbf{X}[s,k]) = \|\mathbf{Y}[s,k] - \widehat{\mathbf{H}}[s,k]\mathbf{X}[s,k]\|^2$$

- Y[s, k] ∈ C<sup>Q×1</sup>: received data of Q antennas at sth OFDM symbol and kth subcarrier
- **Ĥ**[*s*, *k*] ∈ C<sup>Q×U</sup>: estimated FDCHTF matrix (FFT transform of CIRs) at *s*th OFDM symbol and *k*th subcarrier

- X[s, k] ∈ C<sup>U×1</sup>: U users' transmitted data at sth OFDM symbol and kth subcarrier
- ML MUD is NP-hard, and we use discrete-binary evolutionary algorithm to solve this optimisation

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## Turbo SISO MUD-Decoder



- Given estimated channel, EA assisted MUD detects data, which are converted into soft bits
- SISO MUD and SISO channel decoder exchange extrinsic information lite times to enhance decoded bits
- After convergence of turbo detection-decoding, detected bits are remodulated and passed to channel estimator

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## ML Channel Estimation

Given estimated data, ML CE solution

$$\widehat{\mathbf{h}}_q[s] = rg \Big\{ \min_{\mathbf{h}_q[s]} J_{ce}ig(\mathbf{h}_q[s]ig) \Big\}, \ 1 \leq q \leq Q$$

with CE optimisation cost function

$$J_{ce}(\mathbf{h}_{q}[s]) = \|\mathbf{Y}_{q}[s] - \widehat{\mathbf{X}}^{\mathrm{T}}[s]\overline{\mathbf{F}}\mathbf{h}_{q}[s]\|^{2}$$

As  $\mathbf{h}_q[s] \in \mathbb{C}^{UL_{cir} \times 1}$ , search space for each optimisation is a continuous  $(2UL_{cir})$ -dimensional space

- We use a continuous EA to solve this optimisation
  - Continuous EA assisted **channel estimator** and discrete-binary EA aided **turbo MUD-decoder** iterates *I*<sub>ce</sub> times
  - Continuous as well as discrete-binary GA, RWBS, PSO and DEA are detailed in the paper

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## Simulated Multiuser OFDM System

#### Simulation parameters of the multi-user OFDM system

Encoder	Туре	RSC
	Code rate	1/2
	Constraint length	3
	Polynomial	$(g_0,g_1)=(7,5)$
Channel Number of paths <i>L<sub>cir</sub></i>		4
	Path delays	{0,1,2,3}
	Average path gains	{0, -5, -10, -15} (dB)
	Taps: frame to frame	Complex white Gaussian
	Taps: within frame	fading rate $F_D = 10^{-7}$
System	MSs U	4
	Receiver antennas Q	3
	Modulation	16-QAM
	Subcarriers K	64
	Cyclic prefix K <sub>cp</sub>	16

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## **EA Algorithmic Parameters**

#### Algorithmic parameters for EA assisted joint CE and MUD-decoder

CE	Parameter	Value	MUD	Parameter	Value
CGA	Population size Ps	100	DBGA	Population size Ps	100
	Selection ratio rs	0.5		Selection ratio rs	0.5
	Mutation parameter $\gamma$	0.01		Mutation probabi. Mb	0.15
	Mutation probabi. Mb	0.2			
CRWBS	Population size Ps	100	DBRWBS	Population size Ps	100
	Mutation parameter $\gamma$	0.001		Mutation probabi. Mb	0.5
	WBS T <sub>wbs</sub>	40		WBS T <sub>wbs</sub>	40
CPSO	Population size Ps	100	DBPSO	Population size Ps	100
	Cognition learning c <sub>1</sub>	2		Cognition learning c <sub>1</sub>	0.1
	Social learning c2	2		Social learning c2	0.3
CDEA	Population size Ps	100	DBDEA	Population size Ps	100
	Greedy factor p	0.1		Greedy factor p	0.7
	Adaptive factor c	0.1		Adaptive factor c	0.8

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#### **Evaluating Metrics**

- CEA-based CE: given perfect data, and no channel noise ( $N_{\rm o} = 0$ )
- Successful run: achieve target  $J_{ce}(\widehat{\mathbf{h}}_{q,G_{\max}^{i},\text{best}}) < 10^{-4}$  within limit of CF-evaluations:  $\overline{N}_{CF-EVs}^{\lim} = P_{s} \cdot G_{\max}^{\lim} = 100 \times 1000$
- Evaluate statistics:

$$N_{\text{fai}} = N_{\text{fai}} + 1; N_{CF-EVs}^{\text{ran}} = N_{CF-EVs}^{\text{ran}} + P_s$$
  
end if

end for

Average number of CF evaluations per run

$$\overline{\textit{N}}_{\textit{CF-EVs}}^{\text{tot}} = \left(\textit{N}_{\textit{CF-EVs}}^{\text{suc}} + \textit{N}_{\textit{CF-EVs}}^{\text{fai}}\right) / \textit{N}_{\text{tot}}$$

Average number of CF evaluations per successful run

$$\overline{\textit{N}}_{\textit{CF}-\textit{EVs}}^{\rm suc} = \textit{N}_{\textit{CF}-\textit{EVs}}^{\rm suc} / \textit{N}_{\rm suc}$$

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### Evaluating Metrics (continue)

• Efficiency is quantified by normalised average number of CF evaluations per run

$$\overline{R}_{CF-EVs}^{\text{tot}} = \overline{N}_{CF-EVs}^{\text{tot}} / \overline{N}_{CF-EVs}^{\text{lim}}$$

or normalised average number of CF evaluations per successful run

$$\overline{R}_{CF-EVs}^{\rm suc} = \overline{N}_{CF-EVs}^{\rm suc} / \overline{N}_{CF-EVs}^{\rm lim}$$

Smaller R<sup>tot</sup><sub>CF-EVs</sub> or R<sup>suc</sup><sub>CF-EVs</sub>, more efficient CEA-CE
 Reliability of CEA aided channel estimator is measured by failure ratio

$$R_{\rm fai} = N_{\rm fai}/N_{\rm tot}$$

• Similar procedure evaluates efficiency and reliability of DBEA-based MUD, by setting  $G_{\text{max}}^{\text{lim}} = 500$  and  $\overline{N}_{CF-EVs}^{\text{lim}} = M^U = 16^4$ 

• Given perfect CSI, no turbo iterations ( $l_{ite} = 1$ ), and a successful detection run:

 $\mathsf{BER} \to \mathsf{0} \text{ for } G^{\mathrm{run}}_{\mathrm{max}} \leq G^{\mathrm{lim}}_{\mathrm{max}}$ 

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## Efficiency and Reliability



(a) Histograms of efficiency and reliability measures, in terms of  $\overline{R}_{CF-EVs}^{tot}$ ,  $\overline{R}_{CF-EVs}^{suc}$  and  $R_{fai}$ , for four CEA assisted CE schemes

CDEA-CE is best, CRWBS-CE close second, and CGA-CE worst

(b) Histograms of efficiency and reliability measures, in terms of R<sup>tot</sup><sub>CF-EVs</sub>, R<sup>suc</sup><sub>CF-EVs</sub> and R<sub>fai</sub>, for four DBEA assisted MUDs DBGA-MUD is best, DBDEA-MUD close second, and DBPSO-MUD worst

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#### Performance



- (a) CEMSE as function of channel SNR for four EA assisted iterative CE and turbo MUD-decoder schemes
- (b) BER as function of channel SNR for four EA assisted iterative CE and turbo MUD-decoder schemes
  - $l_{\text{ite}} = 3$ ,  $l_{\text{ce}} = 5$ , number of CF evaluations for EA aided CE set to  $N_{CF-EV_S}^{oe} = 20000$  ( $G_{\text{max}} = 200$ ), and number of CF evaluations for EA aided MUD-decoder set to  $N_{CF-EV_S}^{out} = 10000$  ( $G_{\text{max}} = 100$ )

GA based and PSO based schemes do not converge

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#### Performance (continue)



- (a) CE MSE as function of channel SNR for four EA assisted iterative CE and turbo MUD-decoder schemes
- (b) BER as function of channel SNR for four EA assisted iterative CE and turbo MUD-decoder schemes
  - $l_{\text{ite}} = 3$ ,  $l_{\text{ce}} = 5$ , number of CF evaluations for EA aided CE set to  $N_{CF-EVs}^{ce} = 40000$  ( $G_{\text{max}} = 400$ ), and number of CF evaluations for EA aided MUD-decoder set to  $N_{CF-EVs}^{mud} = 20000$  ( $G_{\text{max}} = 200$ )
    - All four schemes converge to optimal solution

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## **Complexity Comparison**

Scheme	Operation	$C_{ m MUD}^{EA}/C_{ m MUD}^{ML}$	$C_{ m turbo}^{EA}/C_{ m turbo}^{ML}$	$C_{\rm joint}^{EA}/C_{\rm turbo}^{ML}$
GA aided joint CE and	multiplications	0.10%	5.69%	62.24%
turbo MUD/decoder	additions	0.10%	7.45%	91.41%
RWBS aided joint CE and	multiplications	0.10%	3.00%	31.27%
turbo MUD/decoder	additions	0.10%	3.88%	45.86%
PSO aided joint CE and	multiplications	0.10%	5.69%	62.24%
turbo MUD/decoder	additions	0.10%	7.45%	91.41%
DE aided joint CE and	multiplications	0.10%	3.00%	31.27%
turbo MUD/decoder	additions	0.10%	3.88%	45.86%

- C<sup>ML</sup><sub>MUD</sub>: complexity of ML MUD given CSI
- C<sup>EA</sup><sub>MUD</sub>: complexity of discrete-binary EA based MUD given CSI
- C<sup>ML</sup><sub>turbo</sub>: complexity of turbo ML MUD-decoder given CSI
- C<sup>EA</sup><sub>turbo</sub>: complexity of discrete-binary EA based turbo MUD-decoder given CSI
- C<sup>EA</sup><sub>joint</sub>: complexity of EA assisted joint CE and turbo MUD-decoder

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## Summary

- Joint channel estimation and turbo multiuser detection-decoding for OFDM communication offers a challenging application
  - to test capabilities of evolutionary algorithms
- Our EA aided joint CE and turbo MUD-decoder is capable of
  - approaching CRLB of optimal channel estimate, and BER of turbo ML MUD-decoder associated with perfect CSI
  - only imposing a fraction of complexity of idealised turbo ML MUD-decoder
- Our study has provided benchmark empirical results to support capabilities of EAs
  - for finding optimal or near optimal designs in challenging practical applications with affordable complexity
  - complimenting well current efforts to better understand EAs