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Nonlinear Beamforming for Multiple-Antenna Assisted QPSK Wireless Systems

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- □ Existing linear beamforming techniques, and motivations for **nonlinear** beamforming or detection
- □ Signal model and optimal Bayesian detection with an inherent **symmetry** property for QPSK beamforming
- □ Symmetric radial basis function network for nonlinear beamforming and adaptive **clustering** algorithm

 $\hfill\square$ Simulation investigation, and performance comparison



Motivations



- □ Classical beamforming is **linear** with a **beampattern** interpretation of beamformer's weight vector:
 - O maximise response at desired user **di**rection and place nulls at interferers' directions, **must** $L \ge M$
 - O similar to **zero-forcing** equalisation, and suffers from **noise enhancement**
- Standard linear beamforming is minimum mean square error (L-MMSE):
 - D better balances nullifying interferers and enhancing noise



- □ State-of-the-art for linear beamforming is **minimum bit error rate** (L-MBER) technique, and in comparison with L-MMSE it offers
 - O Better system BER performance, and larger user capacity
- □ Beamforming can be viewed as **classification**, which classifies received channel-impaired signal into most-likely transmitted symbol point
- $\hfill \Box$ In comparison with linear beamforming, **nonlinear** detection offers
 - **O** significantly better BER performance and much larger user capacity, at cost of higher complexity
- □ With **posterior** or **conditional probabilities** as **generalised beampattern** interpretation
 - O This nonlinear detection can be viewed as **nonlinear beamforming**



- \square *M* single-transmit-antenna users transmit on same carrier, receiver is equipped with *L*-element **antenna array**, channels are non-dispersive
- \square Received signal vector $\mathbf{x}(k) = [x_1(k) \ x_2(k) \cdots x_L(k)]^T$ is

$$\mathbf{x}(k) = \mathbf{P}\mathbf{b}(k) + \mathbf{n}(k) = \bar{\mathbf{x}}(k) + \mathbf{n}(k)$$

 \square $\mathbf{n}(k) = [n_1(k) \ n_2(k) \cdots n_L(k)]^T$ is noise vector, and system matrix

$$\mathbf{P} = [A_1 \mathbf{s}_1 \ A_2 \mathbf{s}_2 \cdots A_M \mathbf{s}_M]$$

□ \mathbf{s}_i is steering vector of source *i*, A_i is *i*-th non-dispersive channel tap □ User *i* is **desired** user, and transmitted symbol vector $\mathbf{b}(k) = [b_1(k) \ b_2(k) \cdots b_M(k)]^T$ with QPSK symbol set

 $b_m(k) \in \{b^{[1]} = +1+j, b^{[2]} = -1+j, b^{[3]} = -1-j, b^{[4]} = +1-j\}, 1 \le m \le M$



□ Denote $N_b = 4^M$ legitimate sequences of $\mathbf{b}(k)$ as \mathbf{b}_q , $1 \le q \le N_b$ □ Noiseless channel state $\bar{\mathbf{x}}(k)$ takes values from set

$$\bar{\mathbf{x}}(k) \in \mathcal{X} = \{ \bar{\mathbf{x}}_q = \mathbf{P} \mathbf{b}_q, 1 \le q \le N_b \}$$

which can be divided into **four subsets** conditioned on $b_i(k) = b^{[m]}$

$$\mathcal{X}^{[m]} \stackrel{\triangle}{=} \{ \bar{\mathbf{x}}_q^{[m]} \in \mathcal{X}, 1 \le q \le N_{sb} : b_i(k) = b^{[m]} \}, \ 1 \le m \le 4$$

Conditional probabilities of receiving $\mathbf{x}(k)$ given $b_i(k) = b^{[m]}$ are

$$p^{[m]}(\mathbf{x}(k)) = \sum_{q=1}^{N_{sb}} \beta_q e^{-\frac{\|\mathbf{x}(k) - \bar{\mathbf{x}}_q^{[m]}\|^2}{2\sigma_n^2}}, \ 1 \le m \le 4$$

 $N_{sb} = N_b/4 = 4^{M-1}$, noise power is $2\sigma_n^2$ and all priors β_q are equal $\square p^{[m]}(\mathbf{x}(k))$ can be interpreted as **generalised beampatterns**

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Optimal detection strategy is

$$\hat{b}_i(k) = b^{[m^*]}$$
 with $m^* = \arg \max_{1 \le m \le 4} p^{[m]}(\mathbf{x}(k))$

 $\hfill\square$ Define complex-valued Bayesian decision variable

$$y_{\text{Bay}}(k) \stackrel{\triangle}{=} b^{[1]} \cdot p^{[1]}(\mathbf{x}(k)) + b^{[2]} \cdot p^{[2]}(\mathbf{x}(k)) + b^{[3]} \cdot p^{[3]}(\mathbf{x}(k)) + b^{[4]} \cdot p^{[4]}(\mathbf{x}(k))$$

 \Box Optimal **Bayesian** detection is: $\hat{b}_i(k) = \operatorname{sgn}(y_{\operatorname{Bay}}(k))$, where

$$\operatorname{sgn}(y) = \begin{cases} b^{[1]} = +1 + j, & y_R \ge 0 \text{ and } y_I \ge 0, \\ b^{[2]} = -1 + j, & y_R < 0 \text{ and } y_I \ge 0, \\ b^{[3]} = -1 - j, & y_R < 0 \text{ and } y_I < 0, \\ b^{[4]} = +1 - j, & y_R \ge 0 \text{ and } y_I < 0, \end{cases}$$

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☐ Four state subsets satisfy following **symmetric** properties

$$\mathcal{X}^{[2]} = +j \cdot \mathcal{X}^{[1]}, \ \mathcal{X}^{[3]} = -1 \cdot \mathcal{X}^{[1]}, \ \mathcal{X}^{[4]} = -j \cdot \mathcal{X}^{[1]}$$

 \Box Thus **Bayesian solution** becomes, for $\bar{\mathbf{x}}_q^{[1]} \in \mathcal{X}^{[1]}$,

$$y_{\text{Bay}}(k) = \sum_{q=1}^{N_{sb}} \left\{ b^{[1]}\beta \cdot e^{-\frac{\|\mathbf{x}(k) - \bar{\mathbf{x}}_{q}^{[1]}\|^{2}}{2\sigma_{n}^{2}}} + b^{[2]}\beta \cdot e^{-\frac{\|\mathbf{x}(k) - j \cdot \bar{\mathbf{x}}_{q}^{[1]}\|^{2}}{2\sigma_{n}^{2}}} \right. \\ \left. + b^{[3]}\beta \cdot e^{-\frac{\|\mathbf{x}(k) + \bar{\mathbf{x}}_{q}^{[1]}\|^{2}}{2\sigma_{n}^{2}}} + b^{[4]}\beta \cdot e^{-\frac{\|\mathbf{x}(k) + j \cdot \bar{\mathbf{x}}_{q}^{[1]}\|^{2}}{2\sigma_{n}^{2}}} \right\}$$

- □ If system channel matrix P can be estimated, as in uplink, subset $\mathcal{X}^{[1]}$ can be calculated and Bayesian solution is specified
- □ In **downlink**, receiver only has access to desired user's training data, estimating **P** is difficult, and other adaptive means has to be adopted

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□ Consider generic **radial basis function** network

$$y_{\text{RBF}}(k) = \sum_{q=1}^{N_c} \left\{ \alpha_q^{[1]} \varphi(\mathbf{x}(k); \mathbf{c}_q, \sigma_q^2) + \alpha_q^{[2]} \varphi(\mathbf{x}(k); j\mathbf{c}_q, \sigma_q^2) \right\}$$

$$+\alpha_q^{[\sigma]}\varphi(\mathbf{x}(k); -\mathbf{c}_q, \sigma_q^2) + \alpha_q^{[\Gamma]}\varphi(\mathbf{x}(k); -\jmath\mathbf{c}_q, \sigma_q^2) \Big\}$$

is number of RBF units, $\varphi(\bullet)$ is usual RBF function, \mathbf{c}_q RBF ce

- □ N_c is number of RBF units, $\varphi(\bullet)$ is usual RBF function, \mathbf{c}_q RBF centres, and σ_q^2 RBF variances
- □ This RBF network has the same **symmetric property** as the Bayesian detector $y_{\text{Bay}}(k)$
- $\square N_c = N_{sb}, \text{ all } \alpha_q^{[m]} = \beta b^{[m]}, \text{ all } \sigma_q^2 = \hat{\sigma}_n^2: N_{sb} \text{ is usually known, } \beta \text{ is any}$ positive number, and $\hat{\sigma}_n^2$ an estimate of noise variance
- \square One only needs to determine all **centres** \mathbf{c}_q



Clustering

□ membership function

$$\mathcal{M}_{l}(\mathbf{x}) = \begin{cases} 1, & \text{if } \bar{v}_{l} \|\mathbf{x} - \mathbf{c}_{l}\|^{2} \leq \bar{v}_{q} \|\mathbf{x} - \mathbf{c}_{q}\|^{2}, \forall q \neq l, \\ 0, & \text{otherwise,} \end{cases}$$

□ cluster **variation**

$$\bar{v}_l(k) = \mu_v \bar{v}_l(k-1) + (1-\mu_v) \mathcal{M}_l(\check{\mathbf{x}}(k)) \|\check{\mathbf{x}}(k) - \mathbf{c}_l(k-1)\|^2$$

 μ_v slightly less than 1.0, all $\bar{v}_l(0)$ set to same small number



Simulation Example

- □ 2-element array with half wavelength spacing, three equal-power QPSK users with **angles of arrival** \rightarrow user 1: 15°, user 2: -60°, user 3: 45°
- □ Bit error rate comparison of theoretical linear MBER beamforming and nonlinear Bayesian beamforming





- \Box User-one, SNR= 20 dB, average over 10runs
- □ (a) **convergence** of clustering, Euclidean **distance** between RBF centres and true channel states, and (b) **insensitivity** to RBF variance





□ Bit error rate of clustering

RBF beamforming for user one, in comparison with optimal **Bayesian** beamforming based on perfect channel knowledge





- □ Nonlinear beamforming achieves significantly smaller system bit error rate and larger user capacity
- Optimal Bayesian beamforming solution for QPSK has an inherent symmetry structure
- □ A novel symmetric radial basis function network has been proposed for QPSK nonlinear beamforming
- □ An adaptive algorithm for downlink senario: clustervariation enhanced clustering





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