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Joint Maximum Likelihood Channel Estimation and Data Detection for MIMO Systems

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- □ Motivations for joint maximum likelihood channel estimation and data detection for MIMO
- □ MIMO Signal model and proposed semi-blind joint ML channel estimation and data detection
- $\hfill\square$ Simulation investigation and performance comparison



- □ Knowledge of **channel state information** is critical to achieve capacity enhancement promised by MIMO, but perfect CSI is often unavailable
- □ Estimating MIMO channel matrix is a tough job, and **training**-based channel estimation is simple but it reduces achievable throughput
- □ Blind joint channel estimation and data detection does not reduce achievable throughput but is computationally complex
- □ To resolve **ambiguities** in channel estimation and symbol detection, a few pilot symbols, i.e. some training, are necessary
- □ We propose a **semi-blind** joint maximum likelihood channel estimation and data detection scheme



 \Box MIMO system of n_T transmitters/ n_R receivers with flat fading channels

$$\mathbf{y}(k) = \mathbf{H}\mathbf{s}(k) + \mathbf{n}(k)$$

❑ Transmitted symbol vector s(Received signal vector y Channel AWGN vector n(

$$\mathbf{s}(k) = [s_1(k) \ s_2(k) \cdots s_{n_T}(k)]^T$$
$$\mathbf{y}(k) = [y_1(k) \ y_2(k) \cdots y_{n_R}(k)]^T$$
$$\mathbf{n}(k) = [n_1(k) \ n_2(k) \cdots n_{n_R}(k)]^T$$

 $\square n_R \times n_T \text{ channel matrix } \mathbf{H} \text{ with } \mathbf{H}(p,m) = h_{p,m}, \text{ for } 1 \leq p \leq n_R \text{ and}$ $1 \leq m \leq n_T$

 \Box $h_{p,m}$ is a complex Gaussian process with zero mean and $E[|h_{p,m}|^2] = 1$

□ Block fading is assumed, where $h_{p,m}$ is kept constant over small block of N symbols



 \Box Define $n_R \times N$ matrix of received data

$$\mathbf{Y} = [\mathbf{y}(1) \ \mathbf{y}(2) \cdots \mathbf{y}(N)]$$

and corresponding $n_T \times N$ matrix of transmitted data

$$\mathbf{S} = [\mathbf{s}(1) \ \mathbf{s}(2) \cdots \mathbf{s}(N)]$$

 \square Knowing data **S**, channel **H** can be estimated by **LSCE**

$$\hat{\mathbf{H}}_{LSCE} = \mathbf{Y}\mathbf{S}^{H}\left(\mathbf{S}\mathbf{S}^{H}\right)^{-1}$$

 \square Knowing channel **H**, **ML detection** of **S** can be performed using OHRSA

J. Akhtman, A. Wolfgang, S. Chen and L. Hanzo, "An optimized-hierarchy-aided approximate Log-MAP detector for MIMO systems," *IEEE Trans. Wireless Communications*, Vol.6, No.5, pp.1900–1909, 2007



□ Both channel and data are **unknown**, joint ML channel and data estimation is defined by

$$(\hat{\mathbf{S}}, \hat{\mathbf{H}}) = \arg \left\{ \min_{\check{\mathbf{S}}, \check{\mathbf{H}}} J_{ML}(\check{\mathbf{S}}, \check{\mathbf{H}}) \right\}$$

where

$$J_{ML}(\check{\mathbf{S}},\check{\mathbf{H}}) = \frac{1}{n_R \times N} \sum_{k=1}^N \left\| \mathbf{y}(k) - \check{\mathbf{H}} \check{\mathbf{s}}(k) \right\|^2$$

but this joint ML search is computationally **prohibitive**

□ Joint optimisation can be decomposed into tractable **iterative loop** first over all possible data and then over all possible channels

$$(\hat{\mathbf{S}}, \hat{\mathbf{H}}) = \arg \left\{ \min_{\check{\mathbf{H}}} \left[\min_{\check{\mathbf{S}}} J_{ML}(\check{\mathbf{S}}, \check{\mathbf{H}}) \right] \right\}$$



 $\Box \quad \textbf{Upper-level Optimisation: RWBS}^{\dagger} \text{ searches MIMO channel space to} \\ \text{find optimal channel estimate } \hat{\mathbf{H}} \text{ by minimising MSE} \\ \end{array}$

$$J_{MSE}(\check{\mathbf{H}}) = J_{ML}(\hat{\mathbf{S}}(\check{\mathbf{H}}), \check{\mathbf{H}})$$

 $\hat{\mathbf{S}}(\check{\mathbf{H}})$ denotes ML estimate of transmitted data for given channel $\check{\mathbf{H}}$

□ Lower-level Optimisation: Given MIMO channel matrix $\check{\mathbf{H}}$, OHRSA detector finds ML estimate of transmitted data $\hat{\mathbf{S}}(\check{\mathbf{H}})$

 \mathbf{A} Feeds back corresponding ML metric $J_{MSE}(\mathbf{\check{H}})$ to upper level

[†]S. Chen, X.X. Wang and C.J. Harris, "Experiments with repeating weighted boosting search for optimization in signal processing applications," *IEEE Trans. Systems, Man and Cybernetics, Part B*, Vol.35, No.4, pp.682–693, 2005



- \Box Pure **blind** joint ML estimation converges slowly and solution $(\hat{\mathbf{S}}, \hat{\mathbf{H}})$ suffers from inherent permutation and scaling **ambiguity** problem
- □ Effective means of resolving ambiguities is to employ a few **pilot symbols** to determine **unitary** $n_T \times n_T$ permutation and scaling matrix
- □ Since we have a few pilots, it is **semi-blind**

 \Box Let number of pilots be t, we can further use **training** data

 $\mathbf{Y}_t = [\mathbf{y}(1) \ \mathbf{y}(2) \cdots \mathbf{y}(t)], \ \mathbf{S}_t = [\mathbf{s}(1) \ \mathbf{s}(2) \cdots \mathbf{s}(t)]$

to provide an initial LSCE $\check{\mathbf{H}}_{LSCE} = \mathbf{Y}_t \mathbf{S}_t^H \left(\mathbf{S}_t \mathbf{S}_t^H \right)^{-1}$ for adding RWBS[†]

[†] RWBS evolves population of channels $\{\check{\mathbf{H}}_{i}^{(g)}\}_{i=1}^{P_{S}}$ over a number of generations $1 \leq g \leq N_{G}$. $\check{\mathbf{H}}_{LSCE}$ is used to initialise the search population

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- $\Box \text{ Algorithm initialisation: } \check{\mathbf{H}}_{\text{best}}^{(0)} = \check{\mathbf{H}}_{LSCE}$
- □ Generation loop: for $(g = 1; g \le N_G; g + +)$ { □ Generation initialisation: $\check{\mathbf{H}}_1^{(g)} = \check{\mathbf{H}}_{\text{best}}^{(g-1)}$

$$\check{\mathbf{H}}_{i}^{(g)} = \check{\mathbf{H}}_{1}^{(g)} + (\mathbf{1} + j\mathbf{1})\eta, \ 2 \le i \le P_{S}$$

 η being random variable uniformly distribution in $[-\gamma, \gamma]$ \Box OHRSA ML detector: $\{\hat{\mathbf{S}}(\check{\mathbf{H}}_{i}^{(g)})\}_{g=1}^{P_{S}}$

- □ Weighted boosting search: for $(l = 1; l \le N_I; l + +)$ { WBS/OHRSA: evolve $\{\check{\mathbf{H}}_i^{(g)}, \hat{\mathbf{S}}(\check{\mathbf{H}}_i^{(g)})\}_{i=1}^{P_S}$
- \Box } End of weighted boosting search
- \Box Solution: $\check{\mathbf{H}}_{\text{best}}^{(g)}$
- □ } End of generation loop

$$\Box \text{ Solution: } \left(\check{\mathbf{H}}_{\text{best}}^{(N_G)}, \hat{\mathbf{S}}(\check{\mathbf{H}}_{\text{best}}^{(N_G)})\right)$$



 \square $n_T = 4$ and $n_R = 4$: 4×4 MIMO system with flat fading channel

- □ Each channel $h_{p,m}$ was complex Gaussian process with zero mean and $E[|h_{p,m}|^2] = 1$, block faded, i.e. kept constant over block of N symbols
- □ Modulation scheme: BPSK, data block: N = 50, pilot symbols: t = 4
- □ Simulation was averaged over 100 runs, **complexity** was determined by number of OHRSA(N) evaluations, n_{ev}
- □ Convergence metrics: MSE $J_{MSE}(\hat{\mathbf{H}}(n_{ev}))$ and MCE $J_{MCE}(\hat{\mathbf{H}}(n_{ev}))$, with

$$J_{MCE}(\hat{\mathbf{H}}(n_{\text{ev}})) = \sum_{m=1}^{n_T} \sum_{p=1}^{n_R} \left| h_{p,m} - \hat{h}_{p,m}(n_{\text{ev}}) \right|^2$$

where $\hat{\mathbf{H}}(n_{ev})$ was channel estimate after n_{ev} OHRSA(N) evaluations



Convergence performance, mean square error and mean channel error, of proposed semi-blind joint ML estimation algorithm, with $\gamma = 0.04$





Influence of algorithmic parameter γ to MCE at 800 OHRSA(N) evaluations, and bit error ratio comparison with $\gamma = 0.04$ for semi-blind scheme



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- An algorithm has been proposed for MIMO semi-blind joint maximum likelihood channel estimation and data detection
- The scheme uses RWBS to search MIMO channel space and OHRSA to provide ML data estimates for channel population
- □ A few pilot symbols are used to resolve ambiguity of blind joint ML estimate and to add RWBS search
- □ Effectiveness of proposed semi-blind joint ML scheme has been demonstrated using simulation



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