

Blind FIR Equalisation for High-Order QAM Signalling

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Blind Equalisation

- Baseband channel model

$$r(k) = \sum_{i=0}^{n-1} a_i s(k-i) + e(k)$$

$a_i = a_{iR} + ja_{iI}$: channel taps; $s(k) = s_R(k) + js_I(k) \in M$ -QAM set:

$$\mathcal{S} = \{s_{il} = (2i - Q - 1) + j(2l - Q - 1), 1 \leq i, l \leq Q\}$$

with $Q = \sqrt{M} = 2^L$ and L an integer; $e(k) = e_R(k) + je_I(k)$: Gaussian white with $E[e_R^2(k)] = E[e_I^2(k)] = \sigma_e^2$

- Equaliser

$$y(k) = \sum_{i=0}^{m-1} w_i r(k-i) = \mathbf{w}^T \mathbf{r}(k)$$

$\mathbf{w}(k) = [w_0 \ w_1 \ \dots \ w_{m-1}]^T$: equaliser weight vector with $w_i = w_{iR} + jw_{iI}$;
 $\mathbf{r}(k) = [r(k) \ r(k-1) \ \dots \ r(k-m+1)]^T$: equaliser input vector

Overview

Low-complexity affordable blind equalisation for high-order QAM

- Constant modulus algorithm
- Concurrent CMA and decision-directed scheme (De Castro *et al*, ICC'2001)
- Multi-stage blind clustering or bootstrap maximum *a posteriori* probability scheme (Chen *et al*, ICC'1993)

Comparative study

Concurrent CMA and DD Scheme

De Castro *et al* (ICC'2001): $\mathbf{w} = \mathbf{w}_c + \mathbf{w}_d$

- CMA equaliser \mathbf{w}_c minimises

$$\bar{J}_{\text{CMA}}(\mathbf{w}) = E \left[(|y(k)|^2 - \Delta_2)^2 \right]$$

$$\left. \begin{aligned} \epsilon(k) &= y(k) (\Delta_2 - |y(k)|^2) \\ \mathbf{w}_c(k+1) &= \mathbf{w}_c(k) + \mu_c \epsilon(k) \mathbf{r}^*(k) \end{aligned} \right\}$$

- DD equaliser \mathbf{w}_d minimises

$$\bar{J}_{\text{DD}}(\mathbf{w}) = \frac{1}{2} E [|Q[y(k)] - y(k)|^2]$$

with quantized equalizer output defined by

$$Q[y(k)] = \arg \min_{s_{il} \in \mathcal{S}} |y(k) - s_{il}|^2$$

DD adaptation follows CMA adaptation:

$$\mathbf{w}_d(k+1) = \mathbf{w}_d(k) + \mu_d \delta(\mathcal{Q}[\tilde{y}(k)] - \mathcal{Q}[y(k)])(\mathcal{Q}[y(k)] - y(k))\mathbf{r}^*(k)$$

where $\mathcal{Q}[\tilde{y}(k)]$ and $\mathcal{Q}[y(k)]$ are equaliser hard decisions after and before CMA adaptation, and indicator function

$$\delta(x) = \begin{cases} 1, & x = 0 + j0 \\ 0, & x \neq 0 + j0 \end{cases}$$

- \mathbf{w}_d is updated only if equaliser hard decisions before and after CMA adaptation are the same
- Hard decision directed adaptation: if decision is wrong \Rightarrow error propagation
- If equaliser hard decisions before and after CMA adaptation are the same, decision is probably right one

Bootstrap MAP equaliser maximises log of the *a posteriori* PDF criterion

$$\bar{J}_{\text{MAP}}(\mathbf{w}) = E[J_{\text{MAP}}(\mathbf{w}, y(k))]$$

where

$$J_{\text{MAP}}(\mathbf{w}, y(k)) = \rho \log(p(\mathbf{w}, y(k)))$$

using stochastic gradient algorithm

$$\mathbf{w}(k+1) = \mathbf{w}(k) + \mu \frac{\partial J_{\text{MAP}}(\mathbf{w}(k), y(k))}{\partial \mathbf{w}}$$

given stochastic gradient

$$\frac{\partial J_{\text{MAP}}(\mathbf{w}(k), y(k))}{\partial \mathbf{w}} = \frac{\sum_{q=1}^Q \sum_{l=1}^Q \exp\left(-\frac{|y(k) - s_{ql}|^2}{2\rho}\right) (s_{ql} - y(k))}{\sum_{q=1}^Q \sum_{l=1}^Q \exp\left(-\frac{|y(k) - s_{ql}|^2}{2\rho}\right)} \mathbf{r}^*(k)$$

- $L = \log_2(M)/2$ stage procedure to achieve minimum complexity

Bootstrap MAP Scheme

After correct equalisation, equaliser output

$$y(k) \approx s(k - k_d) + v(k)$$

with

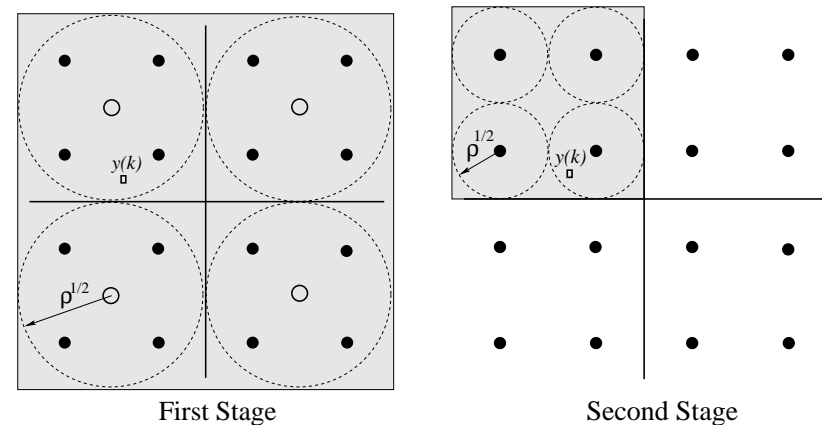
$$\begin{bmatrix} E[v_R^2(k)] & E[v_R(k)v_I(k)] \\ E[v_I(k)v_R(k)] & E[v_I^2(k)] \end{bmatrix} \approx \begin{bmatrix} \rho & 0 \\ 0 & \rho \end{bmatrix}$$

Thus, the *a posteriori* PDF of $y(k)$ is approximately

$$p(\mathbf{w}, y(k)) \approx \sum_{q=1}^Q \sum_{l=1}^Q \frac{p_{ql}}{2\pi\rho} \exp\left(-\frac{|y(k) - s_{ql}|^2}{2\rho}\right)$$

where p_{ql} are the *a priori* probabilities of symbols s_{ql} , $1 \leq q, l \leq Q$.

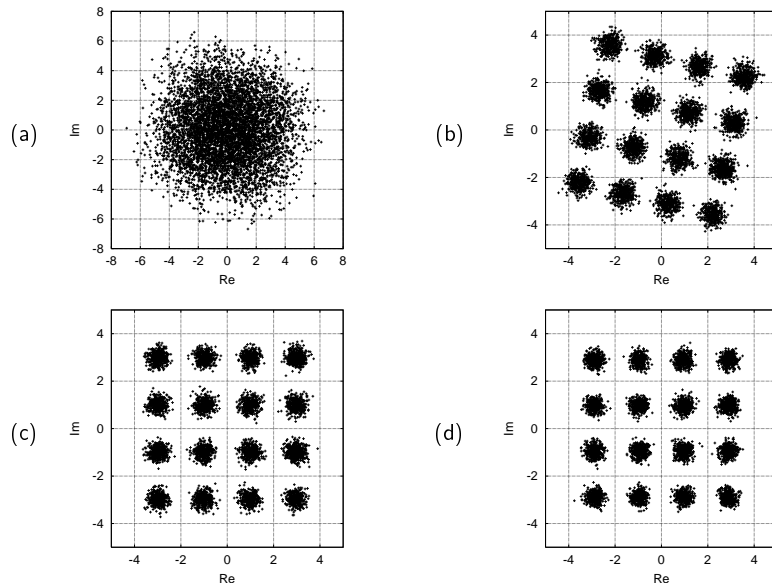
Two-stage for 16-QAM



Computational Complexity per Weight Update

equaliser	multiplications	additions	$\exp(\cdot)$ evaluations
CMA	$8 \times m + 6$	$8 \times m$	—
CMA+DD	$16 \times m + 8$	$20 \times m$	—
MAP	$8 \times m + 23$	$8 \times m + 19$	4

- m is equaliser order
- Four $\exp(\cdot)$ evaluations implemented through look up table
- Tuning of bootstrap MAP equaliser more complicated, as it involves L stage switchings, each having a set of different algorithm parameters



(a) unequalised; (b) CMA; (c) CMA+DD; (d) MAP

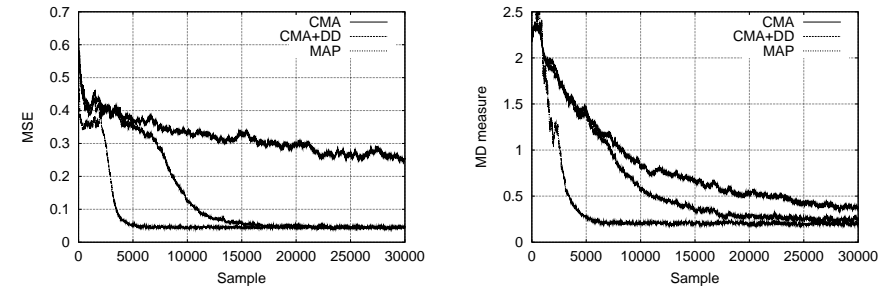
16-QAM Example

22-tap channel, 23-tap equaliser, channel SNR= 25 dB

○ Estimated MSE; and maximum distortion measure:

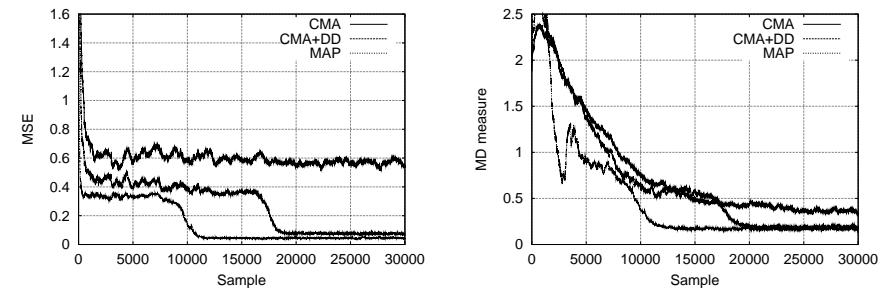
$$MD = \frac{\sum_{i=0}^{n_c-1} |f_i| - |f_{i_{max}}|}{|f_{i_{max}}|}$$

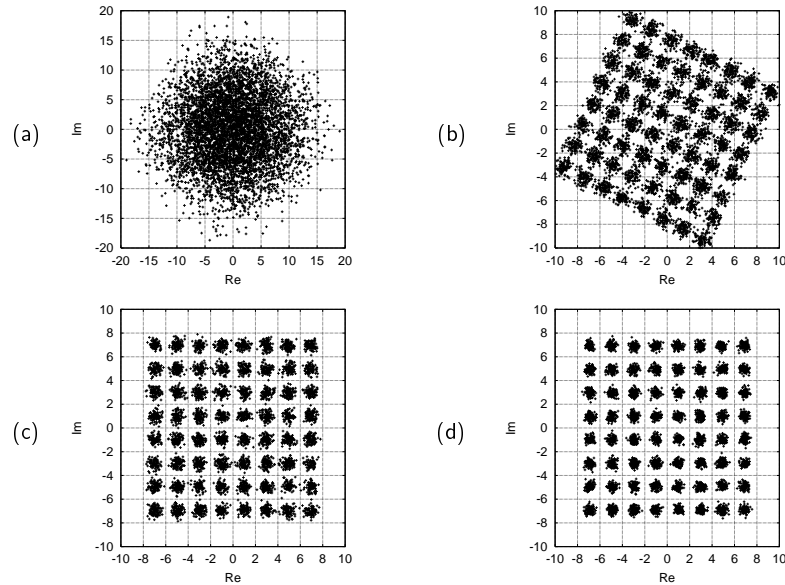
with $\{f_i\}_{i=0}^{n_c-1}$ combined impulse response of channel and equaliser



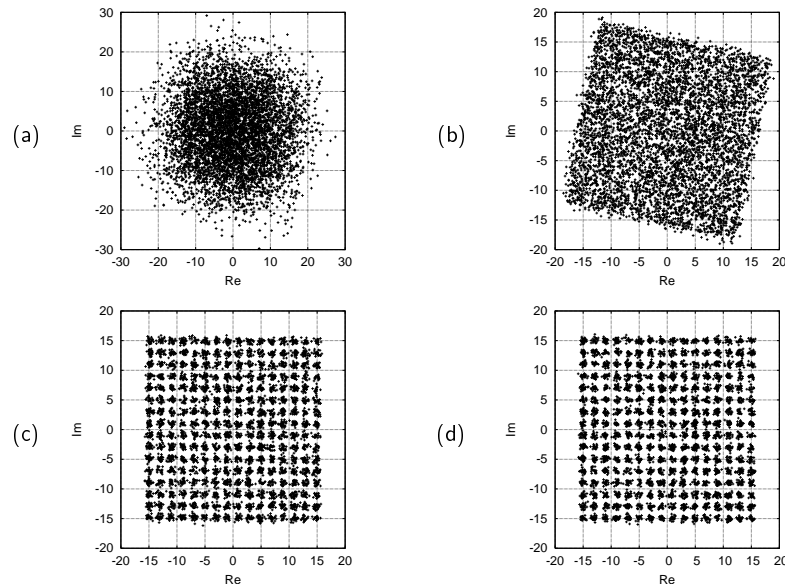
64-QAM Example

5-tap channel, 23-tap equaliser, channel SNR= 30 dB





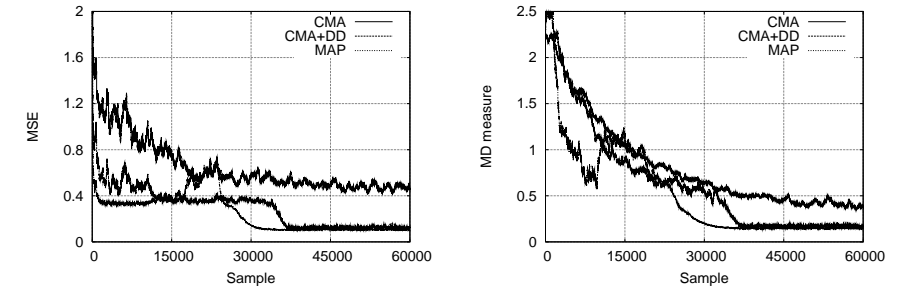
(a) unequalised; (b) CMA; (c) CMA+DD; (d) MAP



(a) unequalised; (b) CMA; (c) CMA+DD; (d) MAP

256-QAM Example

22-tap channel, 23-tap equaliser, channel SNR= 80 dB



Conclusions

Two novel low-complexity blind equalisers: concurrent CMA and DD scheme and bootstrap MAP scheme

↑ Bootstrap MAP blind equaliser has simpler computational complexity per weight update, faster convergence rate, and marginally better steady-state equalisation performance

↓ Tuning of bootstrap MAP blind equaliser is more complicated, as it involves L stage switchings, each having a set of different algorithm parameters

○ Both offers significant improvement in equalisation performance over the very simple CMA